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# Third-Order Robust Fuzzy Sliding Mode Tracking Control of a Double-Acting Electrohydraulic Actuator

Muhammad Fadli Ghani<sup>1</sup>, Rozaimi Ghazali<sup>2</sup>, Hazriq Izzuan Jaafar<sup>3</sup>, Chong Chee Soon<sup>4</sup>, Yahaya Md. Sam<sup>5</sup>, Zulfatman Has<sup>6</sup>

<sup>1</sup>Malaysian Institute of Marine Engineering Technology (MIMET), Universiti Kuala Lumpur, 32200, Lumut, Perak, Malaysia <sup>2,3,4</sup>Centre for Robotics and Industrial Automation, Faculty of Electrical Engineering, Universiti Teknikal Malaysia Melaka, Hang Tuah Jaya, 76100, Durian Tunggal, Melaka, Malaysia

<sup>5</sup>Department of Control and Mechatronics Engineering, School of Electrical Engineering, Universiti Teknologi Malaysia, 81310 Skudai, Johor, Malaysia

<sup>6</sup>Electrical Engineering Department, University of Muhammadiyah Malang, 65144 Malang, Indonesia

Abstract— In the industrial sector, an electrohydraulic actuator (EHA) system is a common technology. This system is often used in applications that demand high force, such as the steel, automotive, and aerospace industries. Furthermore, since most mechanical actuators' performance changes with time, it is considerably more difficult to assure its robustness over time. Therefore, this paper proposed a robust fuzzy sliding mode proportional derivative (FSMCPD) controller. The sliding mode controller (SMC) is accomplished by utilizing the exponential law and the Lyapunov theorem to ensure closed loop stability. By replacing the fuzzy logic control (FLC) function over the signum function, the chattering in the SMC controller has been considerably reduced. By using the sum of absolute errors as the objective function, particle swarm optimization (PSO) was used to optimize the controller parameter gain. The experiment results for trajectory tracking and the robustness test were compared with the sliding mode proportional derivative (SMCPD) controller to demonstrate the performance of the FSMCPD controller. According to the findings of the thorough study, the FSMCPD controller outperforms the SMCPD controller in terms of mean square error (MSE) and robustness index (RI).

Keywords—tracking control, electrohydraulic actuator (EHA), robust control design, fuzzy sliding mode proportional derivative (FSMCPD), particle swarm optimization (PSO).

#### I. INTRODUCTION

Because of their compact size in proportion to their power, strong force producing capabilities, and rapid response times, electrohydraulic actuator (EHA) systems are particularly practical and reliable.

These features make them attractive in construction equipment, sometimes called mobile hydraulic machines, where they are often used [1]-[6]. While the EHA system is known to be a nonlinear system with substantial uncertainties due to the servo valve's high-frequency behavior and external disturbances, it is also known to be a nonlinear system with considerable uncertainties. Furthermore, the presence of uncertainties, nonlinearities, and disturbances in the EHA system usually results in tracking errors and phase lag during the trajectory tracking process, increasing the difficulty of controller design [3], [7]-[14]. These problems make researchers think about how to improve the EHA's performance.

In this work, two types of controllers are principally addressed: a sliding mode proportional derivative (SMCPD) controller and a fuzzy sliding mode proportional derivative (FSMCPD) controller. Because the SMC is only limited by practical restrictions on the amplitude of the control signals it produces, it can theoretically cope with a broad range of uncertainties as well as limited external disturbances. SMC.

The purpose of this research is to present the implementation of FSMCPD controller on a third-order model of a double-acting EHA system. The SMC control has been improved with the addition of a PD sliding surface to increase the performance of the EHA system's trajectory tracking. The Lyapunov criteria are used to demonstrate the stability of the proposed control strategy. Experimenting and comparing the performance of the proposed control methodology to SMCPD controller has shown the utility of the proposed control methodology. The following is a breakdown of the contributions made by the paper:



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a) The FSMCPD controller is presented and successfully demonstrated for effective robust control on a third order model of a double-acting EHA system for trajectory tracking under the influence of supply pressure Consequently, conventional SMCs are typically constructed with a discontinuous switching function. The inaccuracy of the switching operation will result in unpleasant chattering [49]–[57]. During the reaching phase, a fuzzy logic control function is utilized to replace the discontinuous signum function used in conventional variations.

b) Chattering was significantly reduced when the suggested controller's optimized settings were adjusted. This was accomplished by employing the PSO technique to acquire the optimized parameters.

Furthermore, this paper is structured as follows: Section III discusses the current literature. Section III presents the dynamical model of a double-acting EHA system and optimization. Section IV explains how the suggested sliding mode control was developed. Section V presents the experiment findings for the performance of a double-acting electrohydraulic actuator system. Section VI focuses on the performance and potential expansion of the planned control.

# II. LITERATURE REVIEW

Numerous solutions have been suggested to solve the current challenges, with a focus on the establishment of robust control systems to overcome uncertainties, system parameter variation, and disturbances. To cope with the EHA system's complexity and challenges, an intelligent and effective control method will be required. Different types of control approaches have been discovered in the literature during the last decades, in the application of the challenges in the tracking control of EHA systems. A growing number of works on EHA system control have been proposed, ranging from linear control and nonlinear control to intelligent control strategies like generalized predictive control (GPC) [15]-[17], model reference adaptive control (MRAC) [18]-[20], sliding mode control (SMC) [21]–[26], self-tuning fuzzy proportional-integralderivative (PID) [27]-[30], and neural network (NN) [31]-[36]. The SMC control approach has been shown to have high potential and is frequently used in the EHA system.

The SMC, a nonlinear control approach based on variable order models, has been effectively used to the regulation of nonlinear and uncertain systems.

It was widely utilized and extremely successful in the control of nonlinear complex systems. A variety of technical applications, including active suspension systems [37]–[41], pneumatic systems [42]–[47], and active magnetic bearing systems [48], have benefited from the SMC's capabilities.

## III. MODELLING AND OPTIMIZATION

#### A. System Identification

The system identification, which entails modeling the system using sets of input and output data, either with or without previous knowledge of the system's operation. In current work, the model structure estimation is constructed using a mathematical derivation based on the EHA system shown in Fig. 1 and by neglecting non-linearity parameters such as internal or external leakage and valve dynamics [58], the symbols for which are displayed in Table I.

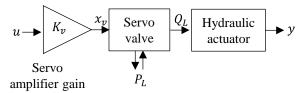


Fig. 1 The EHA system for model structure estimation.

TABLE I
REPRESENTATION OF THE SYMBOLS

Symbol	Representation					
$\overline{u}$	Input signal					
$x_v$	Spool valve position					
$K_v$	Servo valve gain					
$Q_L$	Total oil flow					
$K_q$	Flow-gain coefficient					
$P_L$	Load pressure					
$K_c$	Flow-pressure coefficient					
$eta_e$	Effective bulk modulus					
$V_t$	Total oil volume					
$C_{tp}$	Total leakage coefficient					
$A_p$	Surface area of the piston					
у	Piston position					
$F_a$	Actuator force					
$M_t$	Total mass					



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Eq. (1) defines the association between the input signal, the servo valve gain, and the spool valve position, and Eq. (2) describes the relationship between the overall oil flows dynamics of the EHA system derived from a Taylor series linearization. Furthermore, Eq. (3) indicates the load pressure deviations over time.

$$x_v = K_v u \tag{1}$$

$$Q_L = K_a x_v - K_c P_L \tag{2}$$

$$\dot{P_L} = \frac{4\beta_e}{V_t} (Q_L - C_{tp} P_L - A_p \dot{y}) \tag{3}$$

As a result, the force produced by the actuator from a total mass coupled to the piston's end is characterized as

$$F_a = A_p P_L = M_t \ddot{y} \tag{4}$$

Substituting Eqs. (2) and (3) into the derivative of Eq. (4) produces

$$M_{t}\ddot{y} + M_{t}\ddot{y} \frac{4\beta_{e}}{V_{t}} (K_{c} + C_{tp}) + A_{p}^{2} \frac{4\beta_{e}}{V_{t}} \dot{y}$$

$$= A_{p} \frac{4\beta_{e}}{V_{t}} K_{q} K_{v} u$$
(5)

Let 
$$b_1 = A_p \frac{4\beta_e}{M_t V_t} K_q K_v$$
,  $b_2 = \frac{4\beta_e}{V_t} (K_c + C_{tp})$  and  $b_3 = A_p^2 \frac{4\beta_e}{M_t V_t}$  the Eqn. (5) becomes,

$$\ddot{y} = ua_1 - a_2 \ddot{y} - a_3 \dot{y} \tag{6}$$

Taking a Laplace transform on Eq. (6) becomes

$$\frac{Y(s)}{U(s)} = \frac{b_1}{s(s^2 + b_2 s + b_3)} \tag{7}$$

In this work, the dynamical model of the double-acting EHA system with  $8\times10^3$  kPa supply pressure is developed. The time domain input-output data collected by the experimental hardware and software arrangement presented in Fig. 2 is utilized to determine the continuous transfer function.

The arrangement has a computer unit installed with MATLAB and SIMULINK software; a data acquisition system (DAQ) system with power supply unit, and a hydraulic plant consist of hydraulic power pack, a proportional valve (Bosch Rexroth 4WREE 6 E08-2X/G24K31/A1V), and a hydraulic actuator (Bosch Rexroth – 200 mm single-rod double acting cylinder) attached with wire displacement sensor. Furthermore, the MATLAB/System identification toolbox is utilized to get a transfer function parameter based on Eq. (7) that matches the detailed model by examining the collection of input and output data.

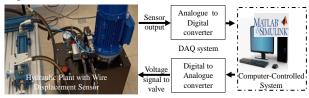


Figure 2. Experimental hardware and software arrangement for system identification.

### B. Particle Swarm Optimization

The PSO can be employed to obtain finest values of the design gain parameters and its algorithm was designed by mimicking the swarm social behaviour of bird flocking and fish schooling. A swarm of individuals called particle move through a high-dimensional search space among the entire population towards the global optimal (minimum or maximum) solution with a specific position and velocity. Each particle in the swarm offers a great solution and adheres to a simple principle by replicating its own previous success [52], [59].

Furthermore, the personal best position in a neighbourhood influences the position of particle and the optimal solution among the personal best positions is known as the global best position. Essentially, the PSO algorithm's implementation may be described as shown in Fig. 3. In the current work, the particle population size, maximum iteration, and cognitive and social coefficients are all set to 20, 50, 2, and 2, respectively.



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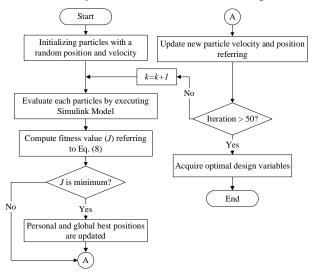


Figure 3. Common process of the PSO algorithm.

$$J = \sum |e(t)| \tag{8}$$

### IV. CONTROLLERS DESIGN

This section describes the design of the SMCPD and FSMCPD controllers. The SMC is a model-based control system in which the output of the controller is built differently for each system under the controller's control. According to the Lyapunov stability theory, the design technique guarantees that the system will persist stable under any conditions [10], [28], [60]–[65].

### A. SMCPD Controllers

In this section the fundamental design of the SMCPD controller is presented. The signal of error, *e* is specified as

$$e = r - y \tag{9}$$

where r is the signal of desired trajectory. Then, the 3rd differentiation of Eq. (9) turns out to be

$$\ddot{e} = \ddot{r} - \ddot{y} \tag{10}$$

In this work, the proportional and derivative sliding surface, s for third order EHA system is considered as

$$s = \left(k_p + k_d \frac{d}{dt}\right)^{n-1} e \tag{11}$$

where n is the order of EHA system, Eq. (11) becomes

$$s = k_p^2 e + 2k_p k_d \dot{e} + k_d^2 \ddot{e}$$
 (12)

Then, the differentiation of Eq. (12) becomes

$$\dot{s} = k_p^2 \dot{e} + 2k_P k_d \ddot{e} + k_d^2 \ddot{e} \tag{13}$$

Combining Eq. (10), Eq. (13) becomes

$$\dot{s} = k_p^2 \dot{e} + 2k_p k_d \ddot{e} + k_d^2 (\ddot{r} - \ddot{y}) \tag{14}$$

With the use of a reaching law, the system output is compelled to follow the surface under consideration. In order to guarantee the stability of the closed loop system, the reaching law must be developed in such a manner that it meets certain criteria. The exponential law [66], as shown in Eq. (15), is used in the proposed investigation.

$$\dot{s} = -\epsilon sgn(s) - ks; \ \epsilon > 0, k > 0 \tag{15}$$

where  $\varepsilon$  and k are constants. Then, solving for Eq. (6), (14) and (15), the control signal of SMCPD is denoted as

$$u_{SMCPD} = \frac{k_p^2}{a_1 k_d^2} \dot{e} + \frac{2k_p}{a_1 k_d} \ddot{e} + \frac{1}{a_1} (\ddot{r} + a_2 \ddot{y} + a_3 \dot{y}) + \frac{1}{a_1 k_d^2} (\epsilon sgn(s) + ks)$$
(16)

#### B. FSMCPD Controller

In this section the vital design of the FSMCPD controller is presented. When the SMC controller is functioned, it causes a fast oscillation phenomenon known as "chattering" at the controller output due to a discontinuity in the signum function that arises during control action. In order to overcome this difficulty, the fuzzy logic control (FLC) denotes as fuzz(s) is employed in the present work instead of the signum function.

$$fuzz(s) > 0; \quad s > 0$$
  
 $fuzz(s) = 0; \quad s = 0$   
 $fuzz(s) < 0; \quad s < 0$  (17)

$$\dot{s} = -\epsilon f u z z(s) - k s; \ \epsilon > 0, k > 0$$

A single-input and single-output FLC has been utilized where the sliding surface is used as input. Three triangular input membership functions (MFs) were used, as shown in Fig. 4, which denoted as Negative Input (NI), Zero Input (ZI), and Positive Input (PI).



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Meanwhile, three output MFs which denoted as Negative Output (NO), Zero Output (ZO), and Positive Output (PO) were used, as shown in Fig. 5. Table II presents the utilized rule base which has been considered from Eq. (17). Mamdani type min-max inference and the center of gravity defuzzification method have been used in the present work.

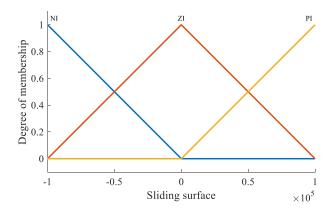


Figure 4. Input membership functions of FLC.

The resulting control action of input output profile is shown in Fig. 6 which implies the rule base design. Then, solving for Eqs. (6), (14) and (17), the control signal of FSMCPD is denoted as

$$u_{FSMCPD} = \frac{k_p^2}{a_1 k_d^2} \dot{e} + \frac{2k_p}{a_1 k_d} \ddot{e} + \frac{1}{a_1} (\ddot{r} + a_2 \ddot{y} + a_3 \dot{y}) + \frac{1}{a_1 k_d^2} (\epsilon f uzz(s) + ks)$$
(18)

TABLE II INPUT-OUTPUT RULE BASE FOR FLC

Input	NI	ZI	PI
Output	NO	ZO	PO

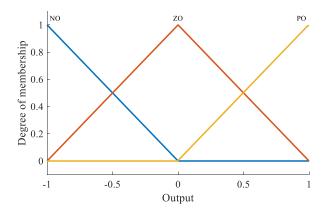


Figure 5. Output membership functions of FLC.

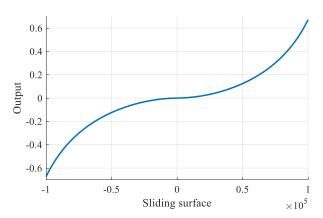


Figure 6. Input-output surface plot of FLC.

#### C. Stability Analysis using Lyapunov Criteria: SMCPD

The crucial aim of SMC controller design is to ensure that the feedback control system is always stable in its overall operation. It is predicted by the Lyapunov stability theorem that when the condition  $s\dot{s} < 0$  is satisfied, the whole system will be stable and will approach the sliding surface. In this present work, the Lyapunov function is denoted as

$$V = \frac{1}{2}s^2 \tag{19}$$

$$\dot{V} = s\dot{s} \tag{20}$$



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Substituting Eq. (15) into Eq. (20) becomes

$$\dot{V} = -s(\epsilon sgn(s) + ks); \ \varepsilon > 0, k > 0 \tag{21}$$

When s > 0 and sgn(s) > 0, Eq. (21) becomes

$$\dot{V} = -s(\epsilon sgn(s) + ks) < 0 \tag{22}$$

When s < 0 and sgn(s) < 0, Eq. (21) becomes

$$\dot{V} = -s(\epsilon sgn(s) + ks) < 0 \tag{23}$$

As revealed by Eq. (22),  $\dot{V} < 0$ , which states that the design controller for the exponential reaching rule will be stable for s > 0. Furthermore, Eq. (23) showed that  $\dot{V} < 0$ , which affirms that the design controller for the exponential reaching rule will be stable for s < 0. Based on Eq. (22) and (23), the Lyapunov stability theory is being used to guide the design of SMCPD controller for third-order doubleacting EHA systems, and the system output will be bound when the input is bounded.

D. Stability Analysis using Lyapunov Criteria: FSMCPD Substituting Eq. (17) into Eq. (20) becomes,

$$\dot{V} = -s(\epsilon f u z z(s) + k s); \ \varepsilon > 0, k > 0 \tag{24}$$

When s > 0 and fuzz(s) > 0, Eq. (21) becomes,

$$\dot{V} = -s(\epsilon f u z z(s) + ks) < 0 \tag{25}$$

When s < 0 and fuzz(s) < 0, Eq. (21) becomes,

$$\dot{V} = -s(\epsilon f u z z(s) + ks) < 0 \tag{26}$$

As revealed by Eq. (25),  $\dot{V} < 0$ , which states that the design controller for the exponential reaching rule will be stable for s > 0. Furthermore, Eq. (26) showed that  $\dot{V} < 0$ , which asserts that the design controller for the exponential reaching rule will be stable for s < 0. Based on Eq. (25) and (26), the Lyapunov stability theory is being used to guide the design of SMCPD controller for third-order doubleacting EHA systems, and the system output will be bound when the input is bounded.

E. Performance Evaluation Criteria

i. Mean Square Error (MSE): MSE disqualifies largevalued errors over small-valued errors, which reflect overshoot and aggressive control. In the current work, the magnitude of chattering phenomenon is reflected relatively proportional to the MSE magnitude.

$$MSE = \frac{1}{n} \sum_{t=0}^{T} (e(t))^{2}$$
 (27)

Robustness Index (RI): The robustness test is practiced assessing the robustness of the established controller. Commonly, robustness test for the EHA system is conducted by reducing or increasing the supply pressure to characterize the parameter variations. The examinations under variations of the operating conditions and robustness study of the employed controllers are vital in control performance evaluation. Accordingly, an applied approach to measure the robustness of the controllers is by determining the MSE for the nominal operating condition (NOM) and under the changed of plant parameters (VAR). In the current work, the robustness and adaptive tests are conducted at 8×10<sup>3</sup> kPa as nominal supply pressure with three variations of supply pressure:  $6 \times 10^3$  kPa as variation 1 (VAR1),  $4\times10^3$  kPa as variation 2 (VAR2), and  $2\times10^3$ kPa as variation 3 (VAR3). The numerical measure which known as the RI for a reference trajectory, under a particular plant condition, and over a tracking process of period is given as

$$RI = \frac{|MSE(NOM) - MSE(VAR)|}{MSE(NOM)}$$
 (28)

# V. EXPERIMENT RESULTS AND DISCUSSION

The assessment of tracking control performance for the SMCPD and FSMCPD controllers are conducted using MATLAB/Simulink (R2021b) with 1 ms of sampling time and optimized parameters using PSO is shown in this section. In addition, the chaotic signal is used to generate the required trajectory. The chaotic trajectory tracking consists of sinusoidal and point-to-point trajectories which is composed together.



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In the experimental study, chaotic trajectory tracking performance of the SMCPD and FSMCPD controllers are assessed which have the fast and slow velocity to accomplish high tracking performance for alteration set points of trajectory throughout the actuator stroke. The values of  $b_1$ ,  $b_2$ , and  $b_3$  are 144400, 3.723, and 7855, respectively, which are developed by the system identification procedure. Furthermore, the values of  $k_p$ ,  $k_d$ ,  $\epsilon$ , and k are 259.2978, 1, 3.0788, and 10.9706, respectively, which are developed by the PSO process. A systematic investigation was carried out, and the results are shown in Table III, where the mean square error (MSE) and robustness index (RI) values are calculated. Meanwhile, in Figs. 7–12, a trajectory tracking task, the related controller output, and error curves are shown, confirming the dominance of the FSMCPD controller over the SMCPD in terms of performance and accuracy.

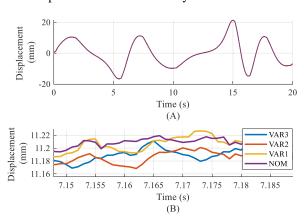


Figure 7. Output performance of SMCPD; (A) Origin and (B) zoom.

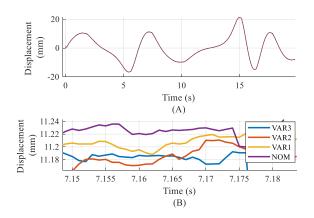


Figure 8. Output performance of FSMCPD; (A) Origin and (B) zoom.

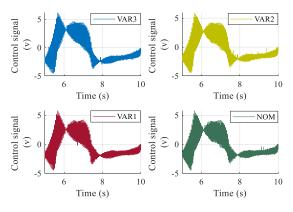


Figure 9. Control signal of SMCPD.

The FSMC delivers more appropriate and improved performance in trajectory tracking control based on MSE analyses, controller efforts, and tracking performance clarifications. In Fig. 13, the sliding surface shows that the system has arrived the sliding phase and will stay there until it reaches the stability point, at which point the error and derivative of error approach zero even in the variation of supply pressure.

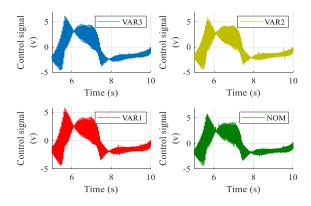


Figure 10. Control signal of FSMCPD.



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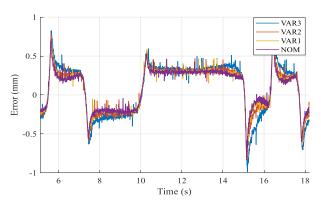


Figure 11. Error signal of SMCPD.

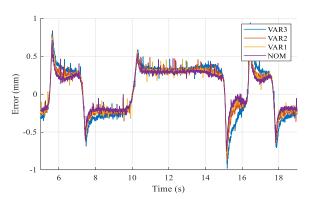


Figure 12. Error signal of FSMCPD.

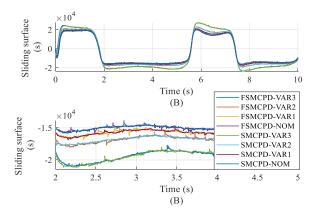


Figure 13. Sliding surface signals; (A) Origin and (B) zoom.

TABLE III
MSE AND RI FOR CHAOTIC TRAJECTORY

CONTROLL ER	MSE (×10 <sup>-2</sup> )			RI (×10 <sup>-2</sup> )			
	NO M	VAR 1	VAR 2	VAR 3	VAR 1	VAR 2	VAR 3
SMCPD	6.86	7.45	8.48	11.43	8.60	23.62	66.62
FSMCPD	6.85	7.43	8.46	11.36	8.47	23.50	65.84

#### VI. CONCLUSION

It is shown in this paper that nonlinear double-acting electrohydraulic actuator (EHA) systems can be controlled using sliding mode proportional derivative (SMCPD) controller, as well as fuzzy sliding mode proportional derivative (FSMCPD) controller. The performance of controllers is evaluated in the context of trajectory tracking tasks and robustness tests, respectively. Chattering is a key problem that must be handled in a traditional sliding mode controller (SMC). This issue is solved by substituting the signum function with a Fuzzy Logic Control (FLC) function. The SMC controllers are built with exponential law which ensuring that the entire system is stable based on Lyapunov's law. The sum of square errors was used to calculate the overall performance index. PSO oversees tuning the gains of the controllers. The experimented results obtained indicate that the FSMCPD controller outperforms the SMCPD controller in all the performance characteristics tested throughout the controller assessment process. It is necessary to test the performance of various SMC versions, such as fractional order sliding mode proportional derivative and fractional order fuzzy sliding mode proportional derivative controllers, which may be explored further based on in this work.

### Acknowledgement

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