

RADIATION EFFICIENCY OF A PERFORATED PANEL MODELLED USING ELEMENTARY SOURCE TECHNIQUE

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A perforate is often used in practical engineering noise control to reduce sound radiation from a vibrating structure. This paper presents a prediction model to quantify this effect using elementary sources which can take account of the hole size and distribution across the plate surface. The model is based on discretization of the Rayleigh formula where a plate set in an infinite rigid baffle is divided into array of elementary source. A matrix of acoustic impedance as a function of acoustic pressure due to motion of the corresponding elementary source and other surrounding sources is assembled. The total sound power is determined from the summation over the power contributions from all elementary sources. The results show that the radiation efficiency reduces as the perforation ratio is increased or the hole size is reduced.

1. Introduction

Perforated plates are frequently found in applications where such structures are used for reducing sound radiation, for examples the safety guard enclosures over flywheels or belt drives and the product collection hoppers. Such perforates are known to be capable of achieving considerable noise reduction. However, until recently, models to calculate their radiated sound are lacking.

Fahy and Thompson¹ developed a model of radiation by plane bending waves propagating in an unbounded plate with uniform perforation which they used to calculate the radiation efficiency of a simply supported rectangular plate. It is assumed that the plate is effectively mounted in a similarly perforated rigid baffle, which limits its usefulness. Takahashi and Tanaka² calculated the radiated power of a one-dimensional perforated panel with infinite length under a point force loading. Putra and Thompson³ proposed a model to calculate the radiation efficiency of a perforated plate by extending Laulagnet's model for radiation from un baffled solid panel⁴ to include the impedance of a distribution of holes.

These established models use the assumption that the array of holes can be replaced by a uniform acoustic impedance at the surface of the plate. Therefore the model gives no information regard-

ing the position of holes and the distance between holes. In this paper, an alternative approach is taken based on elementary sources. In principle, the concept is similar to the models of Vitiello *et al.*⁵ and Cunefare and Koopman⁶ which used discrete elementary radiators to replace a continuous radiator. A continuous baffled source can be modelled by replacing it with an array of point monopole sources. In the same way, a perforated plate set in a rigid baffle can also be modelled by an array of sources, some representing the plate and others the holes.

2. Governing equations

2.1 Discrete version of Rayleigh integral

The well-known Rayleigh integral⁷ can be used to calculate the sound pressure p at any point of observation \mathbf{x} due to a vibrating plate on the assumption that it is set in an infinite rigid baffle. The equation can be written as

$$p(\mathbf{x}) = -2 \int_s G(\mathbf{x}|\mathbf{x}_0) \left(\frac{\partial p(\mathbf{x}_0)}{\partial n} \right) dS \quad (1)$$

where $\mathbf{x} = (x, y, z)$ and S denotes the surface area of the plate (assumed to lie in the xy -plane). A time dependence of $e^{j\omega t}$ is assumed implicitly, where ω is the circular frequency and t is time. G is the free-field Green's function, which is the sound pressure contribution at \mathbf{x} due to a unit point monopole source at \mathbf{x}_0 . This equation can be written as

$$p(\mathbf{x}) = \frac{j\rho ck}{2\pi} \int_s v_p(\mathbf{x}_0) \frac{e^{-jkR}}{R} dS \quad (2)$$

since $\partial p/\partial n = -j\rho ck v_p$ and $G = e^{-jkR}/4\pi R$ where $R = |\mathbf{x} - \mathbf{x}_0|$, v_p is the normal velocity amplitude of the plate, ρ is the air density, c is the speed of sound and $k = \omega/c$ is the acoustic wavenumber.

The Rayleigh integral requires that the normal velocity v_p is known over the whole plate surface area. For bending of a rectangular plate with dimensions $a \times b$ and assuming simply supported edges, the velocity can be written as the sum of modal contributions given by

$$v_p(\mathbf{x}_0) = \sum_{m=1}^{\infty} \sum_{n=1}^{\infty} u_{mn} \sin\left(\frac{m\pi x_0}{a}\right) \sin\left(\frac{n\pi y_0}{b}\right) \quad (3)$$

where u_{mn} is the complex velocity amplitude of mode (m,n) . For a point force excitation at (x_e, y_e) and at frequency ω , it is given by⁸

$$u_{mn} = \frac{4j\omega F}{[\omega_{mn}^2(1 + j\eta) - \omega^2]M} \sin\left(\frac{m\pi x_e}{a}\right) \sin\left(\frac{n\pi y_e}{b}\right) \quad (4)$$

where F is the force amplitude, η is the damping loss factor, M is the plate mass and ω_{mn} is the natural frequency given by

$$\omega_{mn} = \sqrt{\frac{B}{\rho_p h} \left[\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2 \right]} \quad (5)$$

where ρ_p is the plate density, h is the plate thickness and $B = Eh^3/(12(1 - \nu^2))$ is the plate bending stiffness, in which E is Young's modulus and ν is Poisson's ratio. Note that the effect of perforation on the plate bending stiffness and mass has been neglected. Discretizing the Rayleigh integral, Eq. (2) can be re-written as

$$p(\mathbf{x}) = \frac{j\rho ck}{2\pi} \sum_s v_p(\mathbf{x}_s) \frac{e^{-jkR}}{R} dx dy \quad (6)$$

where $R = |\mathbf{x} - \mathbf{x}_s|$ and \mathbf{x}_s is the centre of source position s . However for field positions \mathbf{x} on the surface ($\mathbf{x} = \mathbf{x}_r$) the integrand is singular for $r = s$. To solve the integral for element r , another approximation corresponding to the pressure distribution on the plate surface is needed. Morse and Ingard⁹ give the total force per unit area (pressure) acting on a rectangular piston with small aspect ratio moving with uniform velocity U_n . For a plate of dimensions $a \times b$, where $b \cong a$, the radiation impedance is given by

$$\frac{p}{U_n} = \frac{\rho c k^2}{16} (a^2 + b^2) + \frac{j8\rho c k}{9\pi} \left(\frac{a^2 + ab + b^2}{a + b} \right), \quad ka \ll \pi \quad (7)$$

Applying this to the elemental source dS and assuming for simplicity a square piston, i.e. $a = dx = b = dy$, this reduces to

$$\frac{p}{U_n} = \rho c \left(\frac{k^2 dx^2}{8} + \frac{j4k dx}{3\pi} \right), \quad k dx \ll \pi \quad (8)$$

Combining this with Eq.(6) the Rayleigh integral can be written in the form

$$\mathbf{p} = \mathbf{M}\mathbf{U} \quad (9)$$

where \mathbf{M} is an impedance matrix with elements

$$M_{rs} = \begin{cases} \frac{j\rho c k}{2\pi} \left(\frac{e^{-jkR_{rs}}}{R_{rs}} \right) (dx)^2, & r \neq s \\ \rho c \left(\frac{(k dx)^2}{8} + \frac{j4k dx}{3\pi} \right), & r = s \end{cases} \quad (10)$$

and \mathbf{p} and \mathbf{U} are vectors of pressure and velocity at each element.

2.2 Impedance matrix including perforation

Consider an array of holes on a plate, as shown in cross-section in Fig. 1. Each hole can be considered to be an acoustic source with volume velocity $S_h U_h$ where U_h is the average fluid velocity in the hole and S_h is the area of the hole. The pressure at any point on the plate surface can be written as a sum of the pressure contributions from all sources representing the plate and the holes according to Eq. (9). The matrix \mathbf{M} can be rearranged to give

$$\mathbf{M} = \begin{pmatrix} \cdot & \cdot & \cdot & \cdot \\ \cdot & \mathbf{M}_{p-p} & \cdot & \mathbf{M}_{p-h} \\ \cdot & \cdot & \cdot & \cdot \\ \cdot & \mathbf{M}_{h-p} & \cdot & \mathbf{M}_{h-h} \end{pmatrix} \quad (11)$$

where $p-p$ refers to the pressure at plate locations due to plate sources, $p-h$ refers to the pressure at plate locations due to hole sources, etc. The pressure and velocity can similarly be partitioned into components corresponding to the plate and the holes giving

$$\begin{Bmatrix} \cdot \\ p_p \\ \cdot \\ p_h \end{Bmatrix} = \mathbf{M} \begin{Bmatrix} \cdot \\ U_p \\ \cdot \\ U_h \end{Bmatrix} \quad (12)$$

where p_p and U_p are the acoustic pressure and velocity at the plate location, and p_h and U_h are the same but for the hole location. The pressure at any point on the plate surface can be written as a sum of the pressure contributions from all sources representing the plate and the holes according to Eq. (9).

The matrix \mathbf{M} can be rearranged to give

$$\mathbf{M} = \begin{pmatrix} \cdot & \cdot & \cdot & \cdot \\ \cdot & \mathbf{M}_{p-p} & \cdot & \mathbf{M}_{p-h} \\ \cdot & \cdot & \cdot & \cdot \\ \cdot & \mathbf{M}_{h-p} & \cdot & \mathbf{M}_{h-h} \end{pmatrix} \quad (13)$$

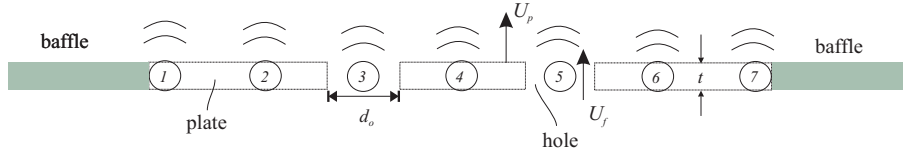


Figure 1. Schematic view of analytical model of an array of discrete (monopole) sources for calculating the sound radiation of a perforated plate

where $p - p$ refers to the pressure at plate locations due to plate sources, $p - h$ refers to the pressure at plate locations due to hole sources, etc. The pressure and velocity can similarly be partitioned into components corresponding to the plate and the holes giving

$$\begin{Bmatrix} -p_p \\ p_h \end{Bmatrix} = \mathbf{M} \begin{Bmatrix} -U_p \\ U_h \end{Bmatrix} \quad (14)$$

where p_p and U_p are the acoustic pressure and velocity at the plate location, and p_h and U_h are the same but for the hole location.

2.3 Impedance of the holes and the acoustic power

It is assumed that the plate is located in an infinite baffle with a semi-infinite halfspace of fluid on both sides. The difference in the local pressure between the upper and lower ends of the hole drives the fluid through the hole (the pressure at the lower surface is $-p_h$). These pressures, in turn, are modified by the fluid flow through the hole.

The sound power radiated by the plate can be expressed as the summation over the power contributions from all discrete elementary sources. Solving the above equations for U_p , the sound power can thus be given by

$$W = \frac{1}{2} \Re \left(\sum p_p U_p^* + \sum p_h U_h^* \right) dS \quad (15)$$

where * indicates complex conjugate.

Finally, the total radiation efficiency of the perforated plate can be written as³

$$\sigma = \frac{W}{\rho c a b (1 - \tau) \langle \overline{v_p^2} \rangle} \quad (16)$$

where τ is the perforation ratio and $\langle \overline{v_p^2} \rangle$ is the spatially-averaged mean square velocity across the total surface of the plate given by

$$\langle \overline{v_p^2} \rangle = \frac{1}{2ab} \sum |U_p|^2 dS \quad (17)$$

Note that the inclusion of $(1 - \tau)$ in Eq. (16) is to normalise the radiated power to the area of the plate. However, it is still assumed here that the bending stiffness of the panel is not affected by the perforation.

3. Accuracy of the result

Firstly, the calculation is made for a solid plate to investigate the accuracy of the result with respect to number of discrete sources employed. The number of the sources is determined by the sample spacing dx and dy . The results are given for a rectangular aluminium plate having dimensions $0.65 \times 0.5 \times 0.003$ m, density $\rho_p = 2700$ kg/m³, Young's modulus $E = 7 \times 10^{10}$ N/m² and damping

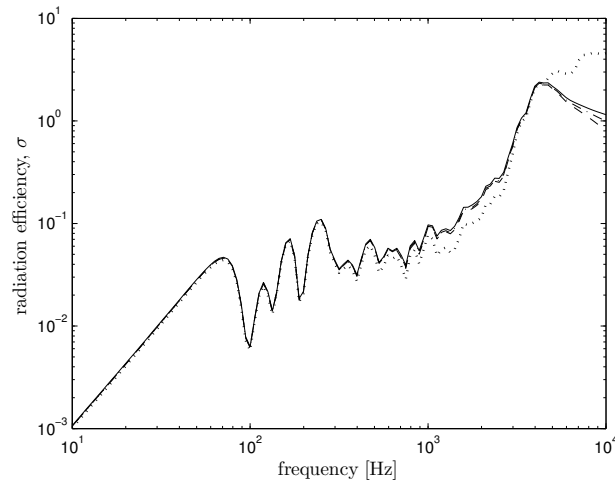


Figure 2. Radiation efficiency of a baffled plate from discrete monopole sources for different grid sizes: \cdots $dx = 40$ mm, $--$ $dx = 20$ mm, $-\cdot-$ $dx = 15$ mm and $—$ $dx = 10$ mm ($0.65 \times 0.5 \times 0.003$ m aluminium plate with $\eta = 0.1$).

loss factor $\eta = 0.1$. The point force is applied at $(-0.13a, -0.1b)$ where the origin $(0,0)$ is at the plate centre. The calculation is performed up to 10 kHz involving all modes with $m \leq 25$ and $n \leq 20$.

The radiation efficiencies for different source spacings (with $dx = dy$) are shown in Fig. 2. As expected, the choice of dx affects the results at high frequency. For $dx = 20$ mm, 15 mm and 10 mm, the curves collapse to one another up to roughly 6 kHz before they separate at higher frequencies. Among these spacings, there are actually small discrepancies below 6 kHz, but these can be considered negligible.

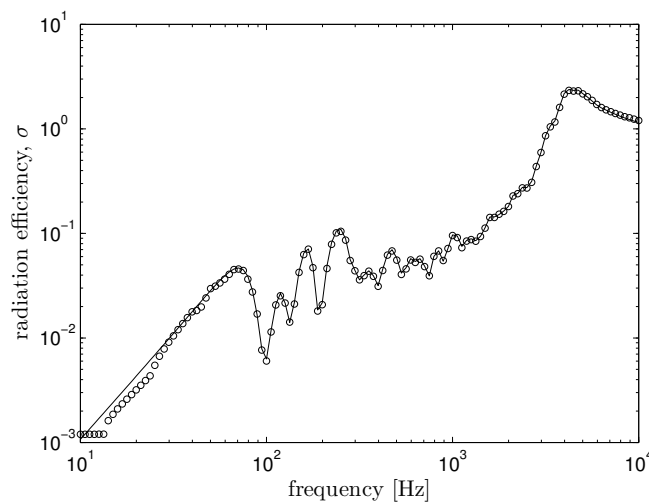


Figure 3. Comparison of radiation efficiency of a baffled plate from discrete monopole sources ($—$) with the FFT approach (\circ)¹⁰ ($0.65 \times 0.5 \times 0.003$ m aluminium plate with $\eta = 0.1$).

Fig. 3 presents a comparison of the radiation efficiency result from Fig. 2 for spacing $dx = 10$ mm with that obtained using the FFT approach¹⁰ for the same plate sample and forcing location. A very good agreement can be seen except at very low frequency, where the result of the FFT approach is known to suffer bias errors.¹⁰ Note that the numerical calculation in the FFT here uses the same sample spacing $dx = dy = 10$ mm. In the remaining sections, the sample spacing $dx = dy = 10$ mm is used in the calculations.

4. The effect of perforation

For the perforated plate, the position of the holes is chosen in relation to the array of the discrete sources. Figs. 4(a) and 4(b) respectively shows examples of the arrangement of holes for constant hole diameter of 40 mm with different hole numbers (different perforation ratios τ) and consequently different hole spacings and for constant perforation ratio of 10% with different hole size to see the effect of hole density (number of holes per unit area). The point force excitation is located at $(-0.054a, -0.08b)$.

The radiation efficiencies are plotted in Fig. 5. Figure 5(a) shows that the level reduces as the perforation ratio increases.³ This is because the counterflow of air through the holes reduces part of the volume sources on the plate surface in the vicinity of the holes, which then reduces the radiated sound. As for the perforated un baffled plate case,³ the sound reduction due to perforation is constant in the fundamental mode region (< 50 Hz). This effect reduces when the frequency approaches the critical frequency. The sound radiation can also be seen to be reduced by decreasing the hole size (i.e. increasing the hole density) as in Fig. 5(b).

In Fig. 5(b), with the smaller hole diameter of 15 mm, a reduction of 15 dB can be achieved at low frequencies with only 10% perforation ratio compared with the corresponding result in Fig. 5(a) with 31% perforation. Hence a high density of small holes is more effective than a low density of large holes.

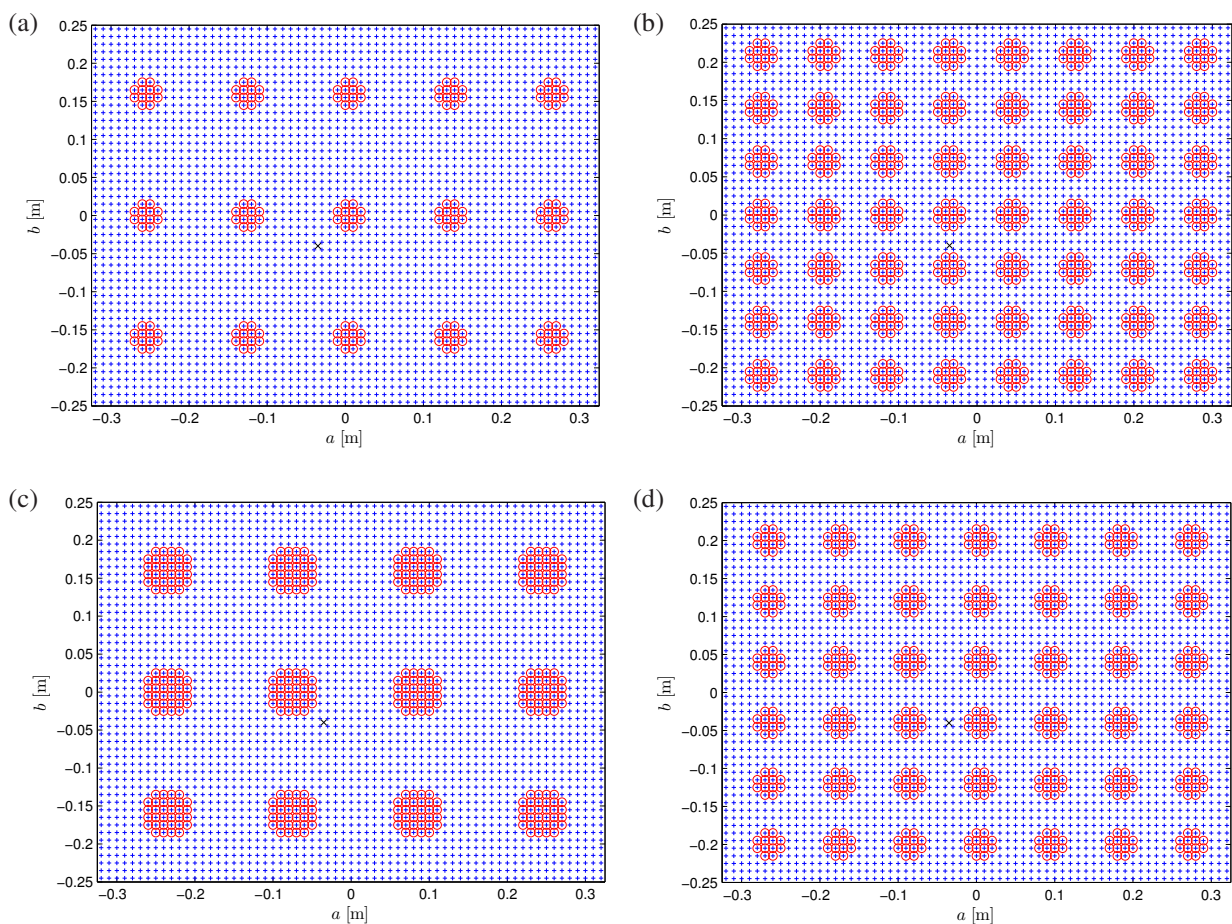


Figure 4. Discrete sources on a rectangular plate (0.65×0.5 m) with arrangement of holes for constant hole diameter $d_o = 40$ mm and different perforation ratios: (a) $\tau = 6\%$, $l_x = 130$ mm, $l_y = 160$; (b) $\tau = 22\%$, $l_x = 80$ mm, $l_y = 70$ mm and for constant perforation ratio $\tau = 10\%$ and different hole diameters: (c) $d_o = 59$ mm, $l_x = 160$ mm, $l_y = 160$ mm; (d) $d_o = 32$ mm, $l_x = 90$ mm, $l_y = 80$ mm (\times : excitation location).

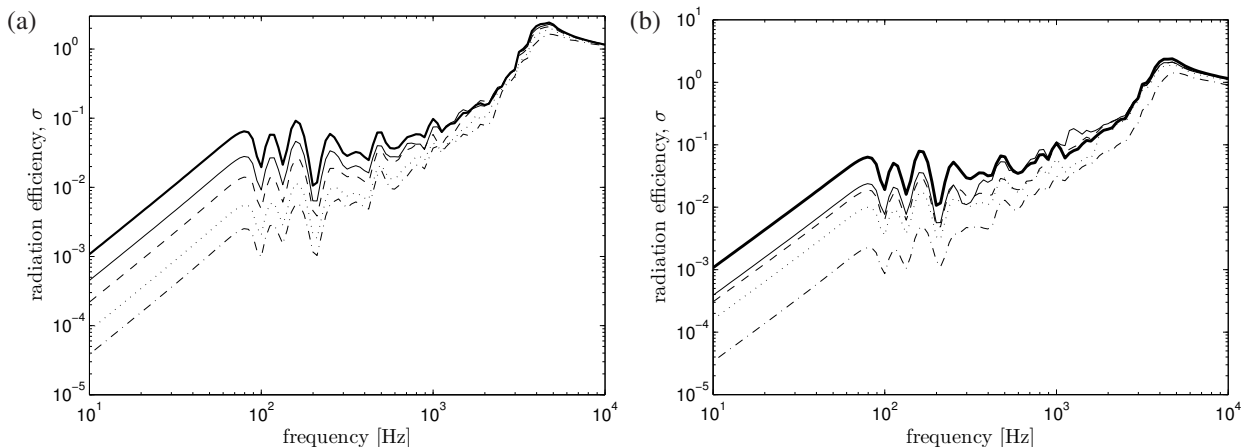


Figure 5. Radiation efficiency of a perforated baffled plate ($0.65 \times 0.5 \times 0.003$ m, $\eta = 0.1$) from discrete monopole sources for (a) constant hole diameter $d_o = 40$ mm: $\text{—}\tau = 6\%$, $\text{--}\tau = 12\%$, $\cdots\tau = 22\%$, $\text{-}\cdot\text{-}\tau = 31\%$ and (b) constant perforation ratio $\tau = 10\%$: $\text{—}d_o = 83$ mm, $\text{--}d_o = 59$ mm, $\cdots d_o = 32$ mm, $\text{-}\cdot\text{-}d_o = 15$ mm (— (thick) unperforated).

5. Conclusions

A model of sound radiation from a perforated baffled plate has been proposed which is developed by dividing the plate and the holes into elementary monopole sources. The contribution of the sources and their influence on each other is assembled into an impedance matrix. It is found that the radiation efficiency reduces as the perforation ratio is increased or the hole size is reduced, consistent with earlier results.

The proposed model avoids the assumption of a continuous layer of impedance as the interaction of the holes with the sound field is analysed individually and the holes are now arranged at a known separation distance. It is therefore of interest to further investigate the effect of hole distance particularly on the frequency limit of sound reduction.

6. Acknowledgment

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